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Connectivity of Underlay Cognitive Radio Networks with Directional Antennas

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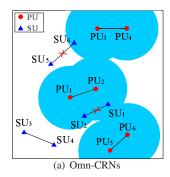
Abstract-In underlay cognitive radio networks (CRNs), the connectivity of secondary users (SUs) is difficult to be guaranteed due to the existence of primary users (PUs). Most prior studies only consider cognitive radio networks equipped with omnidirectional antennas causing high interference at SUs. We name such CRNs with omni-directional antennas as Omn-CRNs. Compared with an omni-directional antenna, a directional antenna can concentrate the transmitting/receiving capability at a certain direction, consequently resulting in the less interference. In this paper, we investigate the connectivity of SUs in CRNs with directional antennas (named as Dir-CRNs). In particular, we derive closed-form expressions of the connectivity of SUs of both Dir-CRNs and Omn-CRNs, thus enabling tractability. We show that the connectivity of SUs is mainly affected by two constraints: the spectrum availability of SUs and the topological connectivity of SUs. Extensive simulations validate the accuracy of our proposed models. Meanwhile, we also show that Dir-CRNs can have higher connectivity than Omn-CRNs mainly due to the lower interference, the higher spectrum availability and the higher topological connectivity brought by directional antennas. Moreover, we also extend our analysis with consideration transmission power efficiency. The simulation results show that Dir-CRNs require less transmission power to establish links than Omn-CRNs. We further investigate the throughput capacity of SUs, which is shown to heavily depend on the connectivity of SUs.

Index Terms—Cognitive Radio Networks, Directional Antennas, Connectivity, Stochastic Geometry, Spectrum Availability.

I. INTRODUCTION

Cognitive radio is a promising technology to improve the efficiency of spectrum usage and to meet the growing demands of high-speed data communications [1]. Cognitive radio networks (CRNs) allow unlicenced secondary users (SUs) to opportunistically access to the spectrum without hampering the communications of licenced primary users (PUs) [2], [3]. CRNs can be roughly categorized into *overlay*, *interweave* and *underlay* paradigms [4]. Overlay CRNs are relatively difficult to be implemented due to the prerequisite of the prior information of PUs [5]. The implementation of the interweave scheme is also challenging because it requires the perfect detection of the existence of PUs, which however is difficult to

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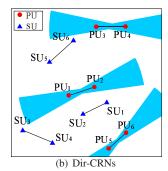


Fig. 1. Omn-CRNs versus Dir-CRNs

be implemented in practice [6]. In an underlay CRN, both SUs and PUs can concurrently use the same spectrum under the provision that the interference caused by SUs at the primary receiver is below a predefined threshold [5], [7]. As a result, a more efficient spectrum utilization can be achieved in underlay CRNs compared with overlay and interweave approaches. Therefore, in this paper we mainly consider underlay CRNs.

A. Related Work and Motivation

In underlay CRNs, connectivity is an important property depicting whether two nodes can establish a communication link. The connectivity of SUs in CRNs is more difficult to be ensured than that of PUs since SUs are susceptible to the existence of PUs having the higher priority to access to the spectrum than SUs. In this paper, we mainly consider the connectivity of SUs. Recently, a number of studies [8]-[13] concentrate on analyzing the connectivity and delay of underlay CRNs. However, most of the studies only consider equipping both PUs and SUs with omni-directional antennas that can cause high interference due to the effect of radiating/receiving signal equally in all directions. We call such CRNs equipped with omni-directional antennas only as *Omn*-CRNs. Take Fig. 1(a) as an example. In this Omn-CRN, only one pair of SUs (i.e., SU₃ and SU₄) can establish a communication link while other SUs such as SU₁, SU₂, SU₅ and SU₆ cannot be active due to the existence of nearby PUs (note that the transmission region of a PU is represented by a shaded circle).

Different from omni-directional antennas, directional antennas can concentrate radio signals on desired directions and consequently reduce the interference. Thus, using directional antennas in CRNs can significantly reduce the interference as indicated in [14]–[16]. However, most recent studies only

consider using directional antennas at either PUs [15] or SUs [14], [16], [17] but not both. The partial deployment of directional antennas in CRNs cannot fully realise the benefits of directional antennas. Therefore, in this paper we propose a novel CRN, in which both PUs and SUs are equipped with directional antennas. We name such CRNs equipped with directional antennas as Dir-CRNs. In a nutshell, Dir-CRNs have the following characteristics: (1) each PU is equipped with a directional antenna; (2) each SU is equipped with a directional antenna; (3) SUs can use the same spectrum as PUs only when the interference caused by SUs is less than a given threshold. Note that we also consider CRNs with SUs equipped with directional antennas for comparison purpose in this paper. We name such CRNs with PUs equipped with omni-directional antennas and SUs equipped with directional antennas as Omn-Dir-CRNs.

Dir-CRNs can potentially improve the connectivity of SUs compared with Omn-CRNs. Take Fig. 1(b) as another example where we consider the same placement of nodes as that in Fig. 1(a) for comparison purpose. In contrast to the Omn-CRN as shown in Fig. 1(a), there are 3 pairs of SUs (i.e., SU_3 and SU_4 , SU_1 and SU_1 , SU_5 and SU_6) that can establish the communication links under the same placement of nodes (note that the transmission range of PUs in Dir-CRNs is longer than that in Omn-CRNs due to the higher antenna gains of directional antennas). This is mainly due to the improvement of spectrum reuse and the inference reduction of directional antennas.

The main goal of this paper is to investigate the connectivity of SUs in Dir-CRNs. It is worth noting that the connectivity of PUs is easier to be guaranteed than that of SUs due to the fact that PUs have higher priority to use the spectrum than SUs in underlay CRNs. To the best of our knowledge, there is no study on the connectivity of SUs in Dir-CRNs. In particular, it is non-trivial to analyse the connectivity of SUs in Dir-CRNs depends on multiple factors, such as the spectrum availability, the topological connectivity, the channel condition and the directivity of antennas.

B. Main Contributions

This paper aims to investigate the connectivity of SUs in both Omn-CRNs and Dir-CRNs. The primary research contributions of this paper can be summarized as follows.

- (1) We formally identify Dir-CRNs that characterize the features of equipping both PUs and SUs with directional antennas. The connectivity of Dir-CRNs has not been studied before.
- (2) We establish a theoretical model to analyse the connectivity of SUs in both Dir-CRNs and Omn-CRNs. In particular, we derive closed-form expressions of the connectivity of SUs in both Dir-CRNs and Omn-CRNs.
- (3) Extensive simulations validate the accuracy of our proposed model. Specifically, our simulations results show that our analytical model can accurately analyse the connectivity of SUs in both Dir-CRNs and Omn-CRNs.
- (4) We also extend the analysis with consideration of the transmission power efficiency. We show that Dir-CRNs

have the higher transmission power efficiency, the higher spectrum availability and the higher topological connectivity. We also analyse the spatial throughput of SUs, which heavily depends on the activities of PUs and the antenna beamwidths of SUs and PUs.

Regarding contribution (1), we identify the characteristics of Dir-CRNs and make a comparison with Omn-CRNs. The benefits of using directional antennas in CRNs will be explained below.

Regarding contribution (2), we derive the connectivity of SUs in both Dir-CRNs and Omn-CRNs based on stochastic geometry. Our model is fairly general since it considers impacts of path-loss attenuation, fading and antenna beamwidth. Interestingly, we find that the connectivity of SUs depends on both the *spectrum availability* and the *topological connectivity*; this result is different from conventional wireless networks, whose connectivity only depends on the topological connectivity.

With respect to contribution (3), we have conducted extensive simulations to verify our proposed model. In particular, our results show that there is an excellent agreement of simulation results with the analytical results, implying the accuracy of our proposed model. Moreover, we also show that Dir-CRNs can significantly improve the connectivity of SUs compared with Omn-CRNs. This connectivity improvement mainly owes to the reduced interference and the improved spectrum utilization by using directional antennas.

Regarding to contribution (4), we extend the analysis with consideration of transmission power efficiency. In particular, we have found that we require less power to establish a link (either a PU link or an SU link) in Dir-CRNs than Omn-CRNs due to the higher antenna gains under the conditions of the same received power and the same transmission range. Moreover, we also extend the analysis to the spatial throughput of SUs, which heavily depends on the connectivity of SUs. Our results show that Dir-CRNs have the higher throughput of SUs than Omn-CRNs.

The rest of the paper is organized as follows. Section II presents system models. We then analyse the connectivity in Section III. Section IV presents the simulation results. We discuss the transmission power efficiency and the throughput of SUs in Section V. Finally, the paper is concluded in Section VI

II. SYSTEM MODELS

A. Network Model

In this paper, we mainly consider two types of CRNs: 1) Omn-CRNs in which both PUs and SUs are equipped with omni-directional antennas; 2) Dir-CRNs in which both PUs and SUs are equipped with directional antennas. Omn-CRNs correspond to conventional cognitive radio wireless networks in which PUs are usually referred to base stations (macrocell base stations) or user equipments (UEs) and SUs are UEs [18]. Recently, there is a new trend of using millimeterwave (mmWave) bands in wireless networks to achieve the extremely high throughput [19], [20]. It becomes feasible to equip both BSs and UEs with directional antennas in mmWave CRNs since the antenna can be quite compact (as the antenna

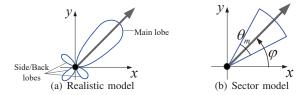


Fig. 2. Directional antennas models

size is inversely proportional to the radio frequency). This type of mmWave CRNs corresponds to the proposed Dir-CRNs. Note that we also consider CRNs with SUs equipped with directional antennas and PUs equipped with omni-directional antennas for comparison purpose. We name this kind of CRNs as Omn-Dir-CRNs, which have been proposed in [14], [16], [17].

In a CRN (either Omn-CRN, Dir-CRN or Omn-Dir-CRN), primary transmitters (PTs) are distributed according to homogeneous Possion point process (PPP) with density λ_p in an infinite two-dimensional Euclidean space. Each PT is associated with a primary receiver (PR), which is randomly and uniformly distributed in the transmission region of PT. According to the displacement theorem [21], the distribution of PRs also follows homogeneous PPP with the same density λ_p . Similarly, secondary users (i.e., SUs including both STs and SRs) are also distributed according to homogeneous PPP with density λ_s . We also assume that SUs are sparsely distributed in the network and there is no overlapping in the communication regions of any two SUs.

B. Channel Model

The radio signal is assumed to undergo both the path loss attenuation and Rayleigh fading [22], [23]. In particular, the path loss attenuation is characterized by the path loss exponent α (usually $2 \le \alpha \le 6$ [24]). Rayleigh fading is modeled as a random variable h following an exponential distribution with mean 1. Let P_t represent the transmitting power. Then, the received power denoted by P_r at a receiver can be expressed as

$$P_r = P_t r^{-\alpha} h G_t G_r. \tag{1}$$

where r is the Euclidean distance between the transmitter and the receiver and G_t and G_r are the antenna gains of the transmitter and the receiver, respectively. We next describe the antenna models as well as antenna gains.

C. Directional Antennas

We introduce the antenna gain to measure the directivity of an antenna. We denote the antenna gain of an omnidirectional antenna by G_o . It is obvious that $G_o=1$ since an omni-directional antenna radiates/receives radio signals uniformly in all directions. Different from omni-directional antennas, directional antennas can concentrate transmitting or receiving capability on desired directions. Generally, a realistic directional antenna typically consists of one or several main beams with the maximum gain and a number of side/backlobes with the relatively lower gains, as shown in Fig. 2(a).

However, the realistic antenna models are so complicated that they are not tractable in theoretic analysis [23], [25], [26]. A sector model is one of typical simplified antenna models [27]–[29], as shown in Fig. 2(b). A sector model consists of one main beam with beamwidth θ_m and all the side/back lobes are ignored. The antenna gain G_d of a sector model is given as follows [27], [28]:

$$G_d(\theta) = \begin{cases} \frac{2\pi}{\theta_m} & \theta \in \left(\varphi - \frac{\theta_m}{2}, \varphi + \frac{\theta_m}{2}\right), \\ 0 & \text{others}, \end{cases}$$
(2)

where $\theta \in [0, 2\pi]$ is the angle from the x-axis in the 2-D coordinate system and φ is the angle of the antenna orientation from x-axis. Note that we usually have the main beam $\theta_m < \pi$.

Eq. (2) is a general expression of the antenna gain $G_d(\theta)$, which mainly depends on the beamwidth θ_m . We consider this sector antenna model in our Dir-CRNs. We denote the antenna beamwidth of PUs and that of SUs by θ_p and θ_s , respectively, and denote the antenna gain of PUs and that of SUs by G_p and G_s , respectively. Note that θ_p is not necessarily equal to θ_s . Replacing θ_m in Eq. (2) by θ_p and θ_s , we then have $G_p = \frac{2\pi}{\theta_p}$ and $G_s = \frac{2\pi}{\theta_s}$. Moreover, we denote the antenna orientation of PTs and that of STs by φ_p and φ_s , respectively. We assume that both φ_p and φ_s follow the uniformly independent identical distribution (i.i.d.) within $[0, 2\pi]$. Once the antenna orientation of a PT or an ST is determined, the corresponding PR or SR can adjust its antenna orientation towards the PT or the ST according to the sophisticated beam-locking schemes [30], [31].

D. Interference constraint

In the underlay spectrum sharing scheme of CRNs, SUs can only access to the spectrum when the interference to PUs (either PTs or PRs) is below an acceptable threshold. In this paper, we implement a *detect-and-avoid* protocol proposed in [32], [33]. This protocol claims that each PR first transmits a detection preamble. If the received power of the preamble at an SU is greater than a threshold η , the SU becomes silent (i.e., it cannot access to the spectrum). In other words, an SU cannot have spectrum if the following condition is satisfied,

$$P_d G_p G_s h R_{ps}^{-\alpha} > \eta, \tag{3}$$

where P_d is the power of detection preamble of PRs (also called the detection power in short), G_p and G_s are the antenna gains of PUs and SUs, respectively, and R_{ps} denotes the distance between a PU and an SU. There are different cases to determine R_{ps} with regard to Dir-CRNs and Omn-CRNs (details will be given in Section III-A). In this manner, SUs can effectively avoid the interference to PUs.

III. CONNECTIVITY

We first define a metric of the connectivity of SUs as the probability of connection, denoted by p_{con} , as follows.

Definition 1: Probability of connection is the probability that any SU pair can successfully establish a bidirectional link.

We then denote the probability of connection of any SU pair in Dir-CRNs, that in Omn-CRNs and that in Omn-Dir-CRNs by p_{con}^d , p_{con}^o and p_{con}^{od} , respectively. Note that we

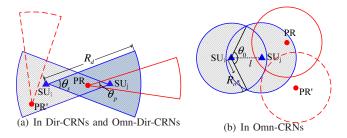


Fig. 3. Detection region of an SU pair

consider bidirectional links since they can guarantee delivering the acknowledgement successfully (e.g., ACK in Wi-Fi). In conventional wireless networks, two nodes can establish a link if they fall into the communication region of each other. We define the condition that any two SUs fall into the communication region of each other¹ as the *topologically-connected condition*. Different from conventional wireless networks, the condition that two SUs in CRNs can establish a link depends on not only the topologically-connected condition [13] but also the condition that the spectrum is available to SUs. We define the condition that the spectrum is available as the *spectrum availability*.

We next define the link condition of SUs as follows.

Definition 2: Link Condition of SUs. An SU pair can successfully establish a link if and only if both the following conditions are satisfied:

- 1) SU pair has spectrum available;
- 2) SU pair is topologically connected with each other.

We denote the event that an SU pair has spectrum available by e_{spe} and the event that an SU pair is topologically connected by e_{top} . It follows that the probability of connection $p_{con}=p(e_{spe}e_{top})=p(e_{spe})p(e_{top}|e_{spe})$, where $p(e_{spe})$ is the probability that an SU pair has spectrum available and $p(e_{top}|e_{spe})$ is the conditional probability that an SU pair can topologically connect to each other under the condition that the SU pair has spectrum available. Therefore, we need to derive the analytical expressions of $p(e_{spe})$ and $p(e_{top}|e_{spe})$ in order to calculate p_{con} . In particular, we analyse $p(e_{spe})$ in Section III-A, then derive $p(e_{top}|e_{spe})$ in Section III-B. Finally, we obtain closed-form expressions of p_{con} in Section III-C.

A. Spectrum Availability

Since there is no overlapping in the communication regions of any two SUs in a sparse network, the analysis on an SU pair applies to any SU pair. We denote the probability that an SU pair (i,j) has spectrum available by p_{ij} . It follows that $p(e_{spe}) = p_{ij}$.

In order to derive p_{ij} , we first give the condition of a pair of SUs having spectrum available. According to the detectand-avoid protocol as described in Section II-D, an SU pair (SU_i and SU_j) has spectrum available if both the following conditions are satisfied:

1) No PR in the detection region of SU_i covers SU_i ;

¹There is an extra condition in Omn-Dir-CRNs and Dir-CRNs: two SUs point their antennas toward each other.

2) No PR in the detection region of SU_i covers SU_i .

The detection region is a region in which SUs cannot receive the detection preamble from PRs. Fig. 3 shows an example of the detection region of an SU pair (SU_i and SU_j). In particular, Fig. 3(a) and Fig. 3(b) show two different cases of Dir-CRNs (or Omn-Dir-CRNs) and Omn-CRNs, respectively. Take Fig. 3(a) as an example, in which the existence of PR' cannot affect SUs because the antenna beamwidth of PR' does not cover SU_i even if PR' falls into the detection region of SU_i . However, there is a different case in Fig. 3(b), which requires no PRs falling into the detection region of SU_i or the detection region of SU_j . Regarding to Omn-Dir-CRNs, we can have the similar case while we only need to consider the antenna direction of SUs since PUs transmit omni-directionally.

As shown in Fig. 3(a), the detection region in Dir-CRNs is the union of two sectors (i.e., the shaded area in Fig. 3(a)), each of which is bounded by the detection range R_d and the beamwidth θ_s . The detection range R_d is the maximum detection distance between an SU and a PR, which can be obtained by letting the left-hand-side (LHS) of Inequality (3) be equal to the right-hand-side (RHS) and replacing the corresponding antenna gains in Inequality (3) by G_p , G_s and $G_o = 1$, respectively.

Different from Dir-CRNs, the detection region of Omn-CRNs is the union of two circles (i.e., the shaded area in Fig. 3(b)), each of which is bounded by the detection range denoted by R_o , which can be obtained by a similar approach to R_d . We then have R_d and R_o as follows, respectively,

$$R_{d} = \left(\frac{4\pi^{2} P_{d} h}{\theta_{p} \theta_{s} \eta}\right)^{\frac{1}{\alpha}},$$

$$R_{o} = \left(\frac{P_{d} h}{\eta}\right)^{\frac{1}{\alpha}}.$$
(4)

We then analyse p_{ij} according to different cases of Dir-CRNs, Omn-CRNs and Omn-Dir-CRNs. We denote the probability that an SU pair has spectrum available in Dir-CRNs by p_{ij}^d and the probability that an SU pair has spectrum available in Omn-CRNs by p_{ij}^o . We first obtain the following result for p_{ij}^d .

Theorem 1: The probability that an SU pair has spectrum available in Dir-CRNs p_{ij}^d is given as follows,

$$p_{ij}^{d} = \exp\left(-\frac{\lambda_p}{2\pi} \left(\frac{4\pi^2 P_d}{\eta}\right)^{\frac{2}{\alpha}} (\theta_p \theta_s)^{\left(1 - \frac{2}{\alpha}\right)} \Gamma\left(1 + \frac{2}{\alpha}\right)\right), \quad (5)$$

where $\Gamma(\cdot)$ is the gamma function.

Proof: Since PRs are randomly distributed according to homogeneous PPP in density λ_p and the antenna direction of each PR is uniformly distributed within $[0,2\pi]$, the event that PRs can cover the SU_i or SU_j follows a homogeneous PPP with density $\lambda_p \cdot \frac{\theta_p}{2\pi}$. Therefore, p_{ij}^d can be expressed as the following equation,

$$p_{ij}^{d} = \left(\frac{\left(\lambda_{p} \cdot \frac{\theta_{p}}{2\pi} \cdot \mathbb{E}[S_{d}]\right)^{0}}{0!} e^{-\lambda_{p} \cdot \frac{\theta_{p}}{2\pi} \cdot \mathbb{E}[S_{d}]}\right)^{2} = \exp\left(-\frac{\lambda_{p} \theta_{p} \mathbb{E}[S_{d}]}{\pi}\right)$$
(6)

where $\mathbb{E}[S_d]$ is the expected value of the area of detection region of an SU in Dir-CRNs (e.g., the shaded region shown

in Fig. 3(a)), which is given by

$$\mathbb{E}[S_d] = \mathbb{E}\left[\pi R_d^2 \cdot \frac{\theta_s}{2\pi}\right] = \frac{\theta_s}{2} \mathbb{E}[R_d^2]. \tag{7}$$

Substituting Eq. (4) into Eq. (7), $\mathbb{E}[S_d]$ can be expressed as follows.

$$\mathbb{E}[S_d] = \frac{\theta_s}{2} \left(\frac{4\pi^2 P_d}{\theta_s \theta_p \eta} \right)^{\frac{2}{\alpha}} \mathbb{E}[h^{\frac{2}{\alpha}}]$$

$$= \frac{\theta_s}{2} \left(\frac{4\pi^2 P_d}{\theta_s \theta_p \eta} \right)^{\frac{2}{\alpha}} \int_0^\infty h^{\frac{2}{\alpha}} e^{-h} dh$$

$$= \frac{\theta_s}{2} \left(\frac{4\pi^2 P_d}{\theta_s \theta_p \eta} \right)^{\frac{2}{\alpha}} \Gamma\left(1 + \frac{2}{\alpha} \right).$$
(8)

Combining Eq. (8) with Eq. (6), we can obtain the results of Eq. (5).

We can extend Theorem 1 to the case of Omn-Dir-CRNs by setting $\theta_p = 2\pi$ in Eq. (5). In particular, we have the following result.

Corollary 1: The probability that an SU pair has spectrum available in Omn-Dir-CRNs p_{ij}^{od} is given as follows,

$$p_{ij}^{od} = \exp\left(-\frac{\lambda_p}{2\pi} \left(\frac{4\pi^2 P_d}{\eta}\right)^{\frac{2}{\alpha}} (2\pi\theta_s)^{\left(1 - \frac{2}{\alpha}\right)} \Gamma\left(1 + \frac{2}{\alpha}\right)\right). \tag{9}$$

Remark 1: Theorem 1 shows that p_{ij}^d depends on multiple factors such as λ_p , η , θ_p , θ_s and α . In particular, p_{ij}^d is decreasing when λ_p increases. In addition, when $\alpha=2$, p_{ij}^d is independent of θ_p and θ_s since $(1-\frac{2}{\alpha})$ becomes 0. However, when $\alpha>2$, p_{ij}^d decreases with increased factor $\theta_p\theta_s$, implying that narrower antenna beamwidth of PUs and SUs brings higher spectrum availability of any SU pair because of fewer SUs that receive detection preambles from PRs. Our numerical results will further confirm this observation.

We then have the result on p_{ij}^o as the following theorem. Theorem 2: The probability that an SU pair can have the spectrum in Omn-CRNs p_{ij}^o is given as follows,

$$p_{ij}^{o} = \exp\left(-\left(\pi + \frac{3\sqrt{3}}{4}\right)\left(\frac{P_d}{\eta}\right)^{\frac{2}{\alpha}}\lambda_p\Gamma\left(1 + \frac{2}{\alpha}\right)\right).$$
 (10)

Proof: Since PRs in Omn-CRNs are equipped with omnidirectional antennas, which can cover all directions, the PRs falling in the detection region of SU_i or SU_j definitely cover SU_i or SU_j . Therefore, if both SU_i and SU_j have spectrum available, there must be no PRs in their detection regions. Then, p_{ij}^o can be expressed as the following equation,

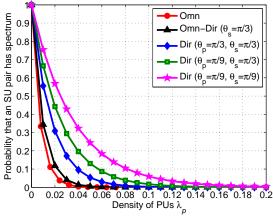
$$p_{ij}^{o} = \frac{(\lambda_p \cdot \mathbb{E}[S_o])^0}{0!} e^{-\lambda_p \cdot \mathbb{E}[S_o]} = \exp\left(-\lambda_p \mathbb{E}[S_o]\right), \quad (11)$$

where $\mathbb{E}[S_o]$ is the expected value of the area of detection region of SU_i union the area of the detection region of SU_j in Omn-CRNs (i.e., the shaded region shown in Fig. 3(b)).

Note that S_o is a random variable depending on the distance between SU_i and SU_j denoted by l (as shown in Fig. 3(b)). Then S_o can be expressed by

$$S_o(l) = (2\pi - \theta_0)R_o^2 + lR_o \sin\frac{\theta_0}{2},$$
 (12)

where $\theta_0 = 2 \arccos \frac{l}{2R_0}$.



(a) $\alpha = 3$

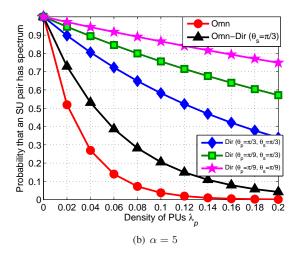


Fig. 4. Probability that any SU pair has spectrum available versus λ_p in Omn-CRNs, Dir-CRNs and Omn-Dir-CRNs (when $\theta_p=2\pi$) with different $F(\theta_s,\theta_p)$ and α , where $P_d=10$.

For simplification, we assume that the maximum value of l is equal to the detection range R_o . According to the fact that each SR is uniformly randomly falling in the transmission region of a ST mentioned in Section II-A, the probability density function (PDF) of the distance l can be expressed as

$$f_l(l) = \frac{2\pi l}{\pi R_o^2} = \frac{2l}{R_o^2} \ (l \le R_o). \tag{13}$$

Then, $\mathbb{E}[S_o]$ can be expressed as

$$\mathbb{E}[S_o] = \mathbb{E}_h(\mathbb{E}_l[S_o]) = \mathbb{E}_h\left(\int_0^{R_o} S_o(l) \cdot f_l(l) dl\right)$$

$$= \int_0^\infty \left(\pi + \frac{3\sqrt{3}}{4}\right) R_o^2 \cdot e^{-h} dh$$

$$= \left(\pi + \frac{3\sqrt{3}}{4}\right) \left(\frac{P_d}{\eta}\right)^{\frac{2}{\alpha}} \Gamma\left(1 + \frac{2}{\alpha}\right).$$
(14)

Combining Eq. (14) with Eq. (11), we obtain the result of Eq. (10).

Fig. 4 plots the analytical results of p_{ij}^d , p_{ij}^{od} and p_{ij}^o with varied density of PUs λ_p under different values of path

loss exponent α , in which Omn, Omn-Dir and Dir represent the results of Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs, respectively. To investigate the impacts of antenna beamwidth on p_{ij}^d , we define $F(\theta_s, \theta_p) = \theta_s \cdot \theta_p$. Specifically, it is shown in Fig. 4 that p_{ij} decreases with the increased value of λ_p , implying that p_{ij} heavily depends on the density of active PUs. In addition, Fig. 4 also indicates that the higher path loss α results in the higher p_{ij} . This is because that the higher path loss results in the smaller detection region. Moreover, in Dir-CRNs, p_{ij}^d decreases when $F(\theta_s, \theta_p)$ decreases from $\frac{\pi^2}{27}$ (green curves) to $\frac{\pi^2}{9}$ (blue curves), but still higher than p_{ij}^{od} and p_{ij}^{o} . This indicates that Dir-CRNs can obtain a higher spectrum availability than Omn-Dir-CRNs and Omn-CRNs. More specifically, Eq. (5) indicates that p_{ij}^d is an increasing function of factor $\theta_s \cdot \theta_p$ (i.e., $F(\theta_s, \theta_p)$) when $\alpha > 2$ (which is quite common in a realistic environment). This observation implies that decreasing the beamwidths of PUs and SUs can improve the spectrum availability, consequently contributing to the connectivity improvement of Dir-CRNs.

B. Topological Connectivity

We next analyse the conditional probability $p(e_{top}|e_{spe})$. To simplify the derivation, we use p_{top} to represent $p(e_{top}|e_{spe})$. We assume that each ST has the same transmission power P_s . The probability of topological connection p_{top} of any SU pair can be expressed as follows,

$$p_{top} = \mathbb{P}[\text{SINR} \geqslant \delta] = \mathbb{P}\left[\frac{P_s r^{-\alpha} h G_s^2}{I_s + I_p + \sigma^2} \geqslant \delta\right],$$
 (15)

where SINR is the signal-to-interference-plus-noise ratio at a reference SU (named as SR_0), δ is the threshold that a link can be established, σ^2 is the thermal noise, I_s and I_p denote the cumulative interference generated from active STs to SR_0 and the cumulative interference from PTs to SR_0 , respectively.

In particular, I_p can be expressed as follows,

$$I_{p} = \sum_{k=1}^{n} P_{p} r_{k}^{\prime - \alpha} h_{k}^{\prime} G_{s_{k}}^{\prime} \widetilde{G}_{p_{k}}^{\prime}, \tag{16}$$

where n denotes the number of PTs, P_p is the transmitting power of each PT, r_k' is the Euclidean distance from PT $_k$ to SR $_0$, h_k' denotes the Rayleigh fading in the channel between PT $_k$ and SR $_j$, G_{s_k}' is the antenna gain of SR $_0$ in the direction of PT $_k$, and \widetilde{G}_{p_k}' is the antenna gain of PT $_k$ in the direction of SR $_0$.

Moreover, I_s can be expressed as follows,

$$I_s = \sum_{g=1}^{m} P_s r_g^{-\alpha} h_g G_{s_g} \widetilde{G}_{s_g}, \tag{17}$$

where m is the number of active STs, P_s is the transmitting power of each ST, r_g is the Euclidean distance from ST $_g$ to SR $_0$, h_g refers to the Rayleigh fading in the channel between PT $_g$ and SR $_0$, G_{s_g} is the antenna gain of SR $_0$ in the direction of ST $_g$, and \widetilde{G}_{s_g} is the antenna gain of ST $_g$ in the direction of SR $_0$.

In order to derive both I_p and I_s , we need to investigate both the distribution of PTs and the distribution of active STs. In particular, the distribution of PTs is affected by the condition that an SU pair has spectrum available. More specifically, Section III-A shows that an SU pair has spectrum available implying no PRs in the detection region of the SU pair. This condition not only restricts the locations of PRs but also confines the locations of PTs due to the correlation between a PT and a PR. Therefore, we can have the approximation that PTs also follow a homogeneous PPP in the whole network under the condition that the SU pair has spectrum available since the detection region of an SU pair is too small compared with the whole network.

With respect to the distribution of STs in Dir-CRNs and Omn-CRNs, we have the following results.

Lemma 1: The active STs of Dir-CRNs follows a thinning homogeneous PPP with density λ_s^d , which can be expressed as the following equation,

$$\lambda_s^d = \lambda_s \cdot p_{ij}^d = \lambda_s \cdot \exp\left(-\frac{\lambda_p}{2\pi} \left(\frac{4\pi^2 P_d}{\eta}\right)^{\frac{2}{\alpha}} (\theta_p \theta_s)^{\left(1 - \frac{2}{\alpha}\right)} \Gamma\left(1 + \frac{2}{\alpha}\right)\right). \tag{18}$$

The active STs of Omn-CRNs also follows a thinning homogeneous PPP with density λ_s^o , which can be expressed as the following equation,

$$\lambda_s^o = \lambda_s \cdot p_{ij}^o = \lambda_s \cdot \exp\left(-\left(\pi + \frac{3\sqrt{3}}{4}\right) \left(\frac{P_d}{\eta}\right)^{\frac{2}{\alpha}} \lambda_p \Gamma\left(1 + \frac{2}{\alpha}\right)\right). \tag{19}$$

Proof: Please refer to Appendix A.

We then have the probability of topological connection of any SU pair, denoted by p_{top}^d as follows.

Theorem 3: The probability of topological connection of any SU pair in Dir-CRNs is given by

$$p_{top}^{d} = \exp\left(-\frac{\delta\sigma^{2}r^{\alpha}\theta_{s}^{2}}{4\pi^{2}P_{s}} - \frac{\delta^{\frac{2}{\alpha}}r^{2}(\lambda_{p}\theta_{s}^{1+\frac{2}{\alpha}}\theta_{p}^{1-\frac{2}{\alpha}}(\frac{P_{p}}{P_{s}})^{\frac{2}{\alpha}} + \lambda_{s}p_{ij}^{d}\theta_{s}^{2})}{2\alpha\sin(\frac{2\pi}{\alpha})}\right),$$
where
$$p_{ij}^{d} = \exp\left(-\frac{\lambda_{p}}{2\pi}\left(\frac{4\pi^{2}P_{d}}{\eta}\right)^{\frac{2}{\alpha}}(\theta_{p}\theta_{s})^{\left(1-\frac{2}{\alpha}\right)}\Gamma\left(1+\frac{2}{\alpha}\right)\right)$$
as given by Eq. (5)).

Proof: The probability of topological connection of any SU pair in Dir-CRNs is

$$p_{top}^{d} = \mathbb{P}\left[\frac{P_{s}r^{-\alpha}hG_{s}^{2}}{I_{s} + I_{p} + \sigma^{2}} \geqslant \delta\right]$$

$$= \mathbb{P}\left[h \geqslant \frac{\delta(I_{s} + I_{p} + \sigma^{2})}{G_{s}G_{s}P_{s}r^{-\alpha}}\right]$$

$$= \exp\left(-\frac{\delta(I_{s} + I_{p} + \sigma^{2})}{G_{s}^{2}P_{s}r^{-\alpha}}\right)$$

$$= \exp\left(-\frac{\delta\sigma^{2}}{G_{s}^{2}P_{s}r^{-\alpha}}\right)\mathcal{L}_{I_{p}}\left(\frac{b}{P_{s}}\right)\mathcal{L}_{I_{s}}\left(\frac{b}{P_{s}}\right),$$
(21)

where $b=\frac{\delta}{G_s^2 r^{-\alpha}}$ and $\mathcal{L}_X(x)$ is the Laplace transform of random variable X at x.

The Laplace transform $\mathcal{L}_{I_p}(\frac{b}{P_s})$ in Eq. (21) can be calculated as follows,

$$\mathcal{L}_{I_p}\left(\frac{b}{P_s}\right) = \mathbb{E}_{I_p}\left[e^{-\frac{bI_p}{P_s}}\right]$$

$$= \mathbb{E}_{h'_k, r'_k, G'_{s_k} \widetilde{G}'_{p_k}}\left[e^{-\frac{bP_p}{P_s} \sum_{k=1}^n r'_k - \alpha h'_k G'_{s_k} \widetilde{G}'_{p_k}}\right]$$

$$= \mathbb{E}_{r'_k, G'_{s_k} \widetilde{G}'_{p_k}} \prod_{k=1}^n \mathbb{E}_h\left[e^{-\frac{bP_p}{P_s} r'_k - \alpha h G'_{s_k} \widetilde{G}'_{p_k}}\right],$$
(22)

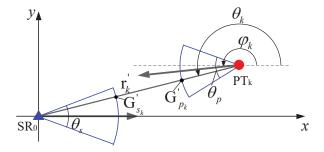


Fig. 5. Schematic illustration of SR₀ and PT_k

where the last step is obtained from the fact that random variables h_k' , r_k' and $G_{s_k}'G_{p_k}'$ are mutually independent and Rayleigh fading variable h_k' is i.i.d. for $1 \le k \le n$.

The probability generation function (PGF) of a PPP (denoted by Φ) in a space S has a property expressed as follows: for a function 0 < f(x) < 1 $(x \in \Phi)$, $\mathbb{E}[\prod_{x \in \Phi} f(x)] = \exp\left(-\lambda \int_S (1-f(x))dx\right)$ [21]. We then use this property and extend it to $R^2 \times [0,2\pi]$ with density $\lambda_p/2\pi$ (with consideration of antenna orientation φ_k) [23]. We next represent Eq. (22) as follows,

$$\mathcal{L}_{I_p}\left(\frac{b}{P_s}\right) = \exp\left(-\frac{\lambda_p}{2\pi} \int_0^{2\pi} \int_{R^2} 1 - \mathbb{E}_h\left[e^{-\frac{bP_p}{P_s} r_k' - \alpha_h G_{s_k}'} \tilde{G}_{p_k}'\right] dr_k' d\varphi_k\right)$$
(23)

where r_k' is the relative position of PT_k in the direction of SR_0 and the expectation $\mathbb{E}_h[e^{-\frac{bP_p}{P_s}r_k'^{-\alpha}hG_{s_k}'}\tilde{G}_{p_k}']$ can be calculated as follows,

$$\begin{split} \mathbb{E}_{h}[e^{-\frac{bP_{p}}{P_{s}}r_{k}'^{-\alpha}hG_{sk}'}\tilde{G}_{pk}'] &= \int_{0}^{\infty}e^{-\frac{bP_{p}}{P_{s}}r_{k}'^{-\alpha}hG_{sk}'}\tilde{G}_{pk}' \cdot e^{-h}dh \\ &= \frac{1}{\frac{bP_{p}}{P_{s}}r_{k}'^{-\alpha}G_{sk}'}\tilde{G}_{pk}' + 1. \end{split}$$

Fig. 5 gives us a schematic illustration of SR_0 and PT_k in a coordinate system, in which both antenna gains G'_{s_k} and G'_{p_k} are determined by φ_k and θ_k , where θ_k denotes the angle from x-axis to the relative direction of SR_0 in the direction of PT_k . Therefore, Eq. (23) can be expressed as follows,

$$\begin{split} &\mathcal{L}_{I_p}\left(\frac{b}{P_s}\right) = \exp\left(-\frac{\lambda_p}{2\pi} \int_0^{2\pi} \int_{R^2}^{\frac{bP_p}{P_s}} r_k'^{-\alpha} G_{s_k}' \widetilde{G}_{p_k}' - G_{k'}' d\varphi_k\right) \\ &= \exp\left(-\frac{\lambda_p}{2\pi} \int_0^{\infty} \int_0^{2\pi} \int_0^{2\pi} \frac{\frac{bP_p}{P_s}}{P_s} r_k'^{-\alpha} G_{s_k}' \widetilde{G}_{p_k}' - G_{k'}' G_{k'}' - G_{k'}' G_{k$$

Following the similar derivation procedure, we have the

$$\mathcal{L}_{I_s}(\frac{b}{P_s})$$
 as follows,

$$\begin{split} &\mathcal{L}_{I_{s}}\left(\frac{b}{P_{s}}\right) = \mathbb{E}_{I_{s}}[e^{-\frac{bI_{s}}{P_{s}}}] = \mathbb{E}_{h_{g},r_{g},G_{sg}}\tilde{G}_{sg}\left[e^{-b\sum_{g=1}^{m}r_{g}^{-\alpha}h_{g}G_{sg}}\tilde{G}_{sg}\right] \\ &= \mathbb{E}_{r_{g},G_{sg}}\tilde{G}_{sg}\prod_{g=1}^{m}\mathbb{E}_{h}[e^{-br_{g}^{-\alpha}hG_{sg}}\tilde{G}_{sg}] \\ &= \exp\left(-\frac{\lambda_{s}p_{ij}^{s}}{2\pi}\int_{0}^{2\pi}\int_{R^{2}}1 - \mathbb{E}_{h}[e^{-br_{g}^{-\alpha}hG_{sg}}\tilde{G}_{sg}]d\mathbf{r}_{g}d\theta_{g}\right) \\ &= \exp\left(-\frac{\lambda_{s}p_{ij}^{s}}{2\pi}\int_{0}^{2\pi}\int_{0}^{2\pi}\int_{0}^{\infty}\frac{br_{g}^{1-\alpha}G_{sg}}{br_{g}^{-\alpha}G_{sg}}\tilde{G}_{sg}}d\phi_{g}d\theta_{g}dr_{g}\right) \\ &= \exp\left(-\frac{\lambda_{s}p_{ij}^{s}(2\pi)^{\frac{4}{\alpha}}b^{\frac{2}{\alpha}}(\theta_{s})^{2-\frac{4}{\alpha}}}{2\alpha\sin\left(\frac{2\pi}{\alpha}\right)}\right), \end{split} \tag{25}$$

where r_g is the relative position of ST_g in the direction of SR_0 , φ_g is the antenna orientation of ST_g and θ_g is the angle from x-axis to the relative direction of SR_0 in the direction of ST_g .

Combining Eq. (25), Eq. (24) and Eq. (21), we have the topological connectivity of any SU pair in Dir-CRNs given by Eq. (20).

Essentially, Theorem 3 is a general expression of p_{top} , which can also be applied for Omn-Dir-CRNs and Omn-CRNs. In particular, we can obtain the probability of topological connection of any SU pair in Omn-Dir-CRNs denoted by p_{top}^{od} by letting $\theta_p = 2\pi$ and replacing p_{ij}^d in in Eq. (20) by p_{ij}^{od} given in Eq. (9). This extension is given by the following corollary.

Corollary 2: The probability of topological connection of any SU pair in Omn-Dir-CRNs is given by

$$p_{top}^{od} = \exp\left(-\frac{\delta\sigma^2 r^\alpha \theta_s^2}{4\pi^2 P_s} - \frac{\delta^{\frac{2}{\alpha}} r^2 (\lambda_p \theta_s^{1+\frac{2}{\alpha}} (2\pi)^{1-\frac{2}{\alpha}} (\frac{P_p}{P_s})^{\frac{2}{\alpha}} + \lambda_s p_{ij}^{od} \theta_s^2)}{2\alpha \sin(\frac{2\pi}{\alpha})}\right),$$
(26)

where p_{ij}^{od} is given by Eq. (9).

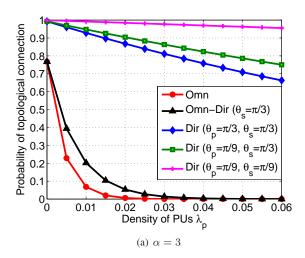
Similarly, we denote the probability of topological connection of any SU pair in Omn-CRNs by p^o_{top} . After replacing $\theta_p=2\pi,\ \theta_s=2\pi$ and p^d_{ij} in Eq. (20) by p^o_{ij} given in Eq. (10), we have the following corollary.

Corollary 3: The probability of topological connection of any SU pair in Omn-CRNs is

$$p_{top}^{o} = \exp\left(-\frac{\delta\sigma^{2}r^{\alpha}}{P_{s}} - \frac{\delta^{\frac{2}{\alpha}}r^{2}\left(\lambda_{p}\left(\frac{P_{p}}{P_{s}}\right)^{\frac{2}{\alpha}} + \lambda_{s}p_{ij}^{o}\right)}{\alpha\sin\left(\frac{2\pi}{\alpha}\right)}\right),$$
 where
$$p_{ij}^{o} = \exp\left(-\left(\pi + \frac{3\sqrt{3}}{4}\right)\left(\frac{P_{d}}{\eta}\right)^{\frac{2}{\alpha}}\lambda_{p}\Gamma\left(1 + \frac{2}{\alpha}\right)\right) \text{ as given by Eq. (10).}$$

Fig. 6 presents the analytical results of probability of topological connection of any SU pair versus distance r with $\alpha=3$ and $\alpha=5$, respectively. In particular, Omn, Omn-Dir and Dir represent the results of Omn-CRNs, Omn-Dir-CRNs and Omn-CRNs, respectively. It is shown in Fig. 6 that p_{top}^d of DIR-CRNs is significantly larger than that of Omn-Dir-CRNs and that of Omn-CRNs.

Remark 2: This result implies that using directional antennas in cognitive radio networks can significantly improve



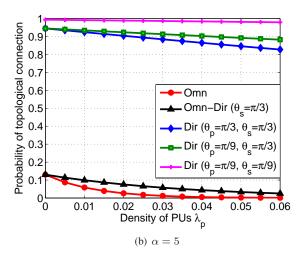


Fig. 6. Probability of topological connection of any SU pair versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with different α , where $P_d=10,\ P_p=8,\ P_s=6,\ r=3,\ \lambda_s=0.0002,\ \eta=0.05,\ \sigma^2=0.01,\ \delta=5.$

$$p_{con}^{d} = p_{top}^{d} \cdot p_{ij}^{d} = \exp\left(-\left(\frac{4\pi^{2}P_{d}}{\theta_{p}\theta_{s}\eta}\right)^{\frac{2}{\alpha}} \frac{\lambda_{p}\theta_{p}\theta_{s}}{2\pi} \Gamma\left(\frac{2+\alpha}{\alpha}\right) - \frac{\delta\sigma^{2}r^{\alpha}\theta_{s}^{2}}{4\pi^{2}P_{s}} - \frac{\delta^{\frac{2}{\alpha}}r^{2}(\lambda_{p}\theta_{s}^{1+\frac{2}{\alpha}}\theta_{p}^{1-\frac{2}{\alpha}}(\frac{P_{p}}{P_{s}})^{\frac{2}{\alpha}} + p_{ij}^{d}\lambda_{s}\theta_{s}^{2})}{2\alpha\sin(\frac{2\pi}{\alpha})}\right). \tag{28}$$

$$p_{con}^{od} = p_{top}^{od} \cdot p_{ij}^{od} = \exp\left(-\left(\frac{2\pi P_d}{\theta_s \eta}\right)^{\frac{2}{\alpha}} \lambda_p \theta_s \Gamma\left(\frac{2+\alpha}{\alpha}\right) - \frac{\delta \sigma^2 r^\alpha \theta_s^2}{4\pi^2 P_s} - \frac{\delta^{\frac{2}{\alpha}} r^2 (\lambda_p \theta_s^{1+\frac{2}{\alpha}} (2\pi)^{1-\frac{2}{\alpha}} (\frac{P_p}{P_s})^{\frac{2}{\alpha}} + p_{ij}^{od} \lambda_s \theta_s^2)}{2\alpha \sin(\frac{2\pi}{\alpha})}\right). \tag{29}$$

$$p_{con}^{o} = p_{top}^{o} \cdot p_{ij}^{o} = \exp\left(-\left(\pi + \frac{3\sqrt{3}}{4}\right)\left(\frac{P_d}{\eta}\right)^{\frac{2}{\alpha}} \lambda_p \Gamma\left(\frac{2+\alpha}{\alpha}\right) - \frac{\delta\sigma^2 r^{\alpha}}{P_s} - \frac{\delta^{\frac{2}{\alpha}} r^2 \left(\lambda_p \left(\frac{P_p}{P_s}\right)^{\frac{2}{\alpha}} + \lambda_s p_{ij}^{o}\right)}{\alpha \sin\left(\frac{2\pi}{\alpha}\right)}\right). \tag{30}$$

the probability of topological connection. This improvement mainly owes to the higher SINR of Dir-CRNs than that of Omn-CRNs. Compared with omni-directional antennas, directional antennas can concentrate the transmission on desired directions while reducing the interference to other undesired directions. Meanwhile, using directional antennas at SUs can also improve the probability of topological connection. This is because using directional antennas at SUs can extend the transmission range, consequently enhancing the probability of topological connection.

We also compare the results with different values of θ_s and θ_p in Dir-CRNs. Fig. 6 shows that p_{top}^d increases when θ_s is fixed at $\frac{\pi}{3}$ and θ_p decreases from $\frac{\pi}{3}$ to $\frac{\pi}{9}$ while p_{top}^d increases when θ_p is fixed at $\frac{\pi}{9}$ and θ_s decreases from $\frac{\pi}{3}$ to $\frac{\pi}{9}$. This result indicates that using narrow-beamwidth antennas at both SUs and PUs can improve the probability of topological connection. Compared with the case of using narrow-beamwidth antennas at SUs can further improve the probability of topological connection.

C. Connectivity

Following the definition of the probability of connection of any SU pair $p_{con}=p(e_{spe}e_{top})=p(e_{spe})p(e_{top}|e_{spe})$ and the derivations of $p(e_{spe})$ and $p(e_{top}|e_{spe})$ in Section III-A and

Section III-B, we then obtain the closed-form expressions of p_{con}^d of Dir-CRNs, p_{con}^{od} of Omn-Dir-CRNs and p_{con}^o of Omn-CRNs as given in Eq. (28), Eq. (29) and Eq. (30), respectively.

We further investigate the impacts of θ_p and θ_s on p_{con}^d of Dir-CRNs. The effect of θ_p and θ_s on p_{con}^d is illustrated in Fig. 7. More specifically, it is shown in Fig. 7 that p_{con}^d increases with the decreased value of θ_p and the decreased value of θ_s , implying that using the narrow-beamwidth antennas at both SUs and PUs improves the connectivity of SUs. However, the narrower antenna beamwidth on the other hand can lead to the challenges in neighbor-discovery [34] and blockage-effect reduction [35]. Therefore, there is a trade-off in choosing the beamwidth. How to choose the appropriate beamwidth at PUs and SUs is one of our future directions.

IV. SIMULATIONS

In this section, we conducted extensive simulations to evaluate the accuracy and the effectiveness of our proposed analytical model on the connectivity of Dir-CRNs, Omn-CRNs and Omn-Dir-CRNs. In particular, we describe the simulation method in Section IV-A. We then present the simulation results in Section IV-B.

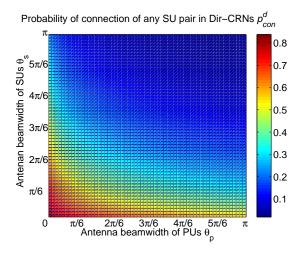


Fig. 7. Probability of connection of an SU pair versus in Dir-CRNs with different θ_p and θ_s , where $P_d=10$, $P_p=8$, $P_s=6$, $\alpha=3$, $\lambda_p=0.02$, $\lambda_s=0.0002$, $\eta=0.05$, $\sigma^2=0.01$, $\delta=5$, r=3.

A. Simulation method

We conduct extensive simulations to evaluate the accuracy of the proposed analytical models. We choose the sophisticated commercial software Matlab as the simulation tool. In particular, PUs are distributed according to HPPP in a plane of area $1,200\times 1,200$. To eliminate the border effect, SUs are distributed according to HPPP on the sub-area with area $1,000\times 1,000$ within the plane [36], [37]. The system parameters are chosen as $P_d=10$, $P_p=8$, $P_s=6$, $\eta=0.05$, $\sigma^2=0.01$, $\delta=5$, $\lambda_s=0.0002$. We denote the simulation result of the probability of connectivity by p_{con}^s in order to differentiate it from the analytical result p_{con} . As shown in [36]–[38], p_{con}^s is given by

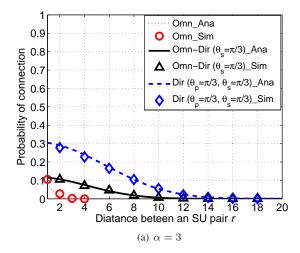
$$p_{con}^{s} = \frac{\text{\# topologies that an SU pair can connect successfully}}{\Omega},$$
(31)

where # means the number of. In order to obtain an approximated result to the analytical one, we need to choose a large enough Ω (theoretically $\Omega \to \infty$) while it is extremely time-consuming to obtain the results of large Ω . In this paper, we choose $\Omega = 3,000$.

B. Simulation results

1) Impact of distance r: In the first set of simulations, we investigate impact of the distance r on the probability of connection of an SU pair in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs. Fig. 8 presents the results, in which the analytical results are plotted as curves and the simulation results are represented as markers. It is shown in Fig. 8 that there is an excellent agreement of the simulation results with the analytical results, implying that our proposed analytical model is fairly accurate.

Meanwhile, it is shown in Fig. 8 that Dir-CRNs has higher value of p_{con}^d than p_{con}^{od} of Omn-Dir-CRNs and p_{con}^o of Omn-CRNs; this implies that using directional antennas instead of omni-directional antennas in CRNs can significantly improve the connectivity of SUs. This improvement mainly owes to the higher spectrum availability and the higher topological



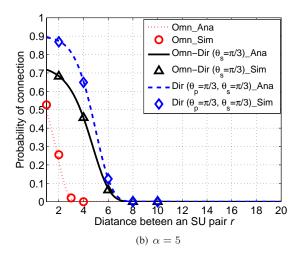
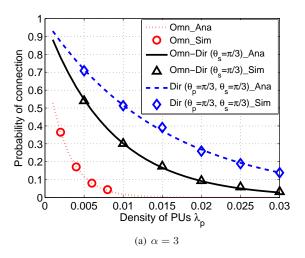


Fig. 8. Probability of connection of an SU pair versus the distance r in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with different values of α , where $\lambda_p=0.02$.

connectivity as indicated in Section III. In addition, aligning Fig. 8(a) with Fig. 8(b) together, we find that p_{con}^d is always higher than p_{con}^{od} and p_{con}^o in when $\alpha=3$ and $\alpha=5$.

- 2) Impact of density of PUs λ_p : In the second set of simulations, we investigate the impact of the density of PUs λ_p . The simulation results are shown in Fig. 9. Similarly, we can observe that the simulation results (represented by markers) match with the analytical results (represented by curves), implying the accuracy of the proposed model. Meanwhile, we also find that the probability of connection of Dir-CRNs, Omn-Dir-CRNs and Omn-CRNs decreases with the increment of node density of PUs; this implies that the activities of PUs have a strong influence on the probability of connection of SUs. Moreover, we also find that Dir-CRNs always outperform Omn-Dir-CRNs and Omn-CRNs in terms of the probability of connection when $\alpha=3$ and $\alpha=5$.
- 3) Impact of beamwidth of PUs and SUs: We next investigate the impact of different beamwidth values of PUs and SUs on the probability of connection in Dir-CRNs. The third set of simulations are conducted to investigate the impact of the beamwith of PUs θ_n . Fig. 10 presents the results where



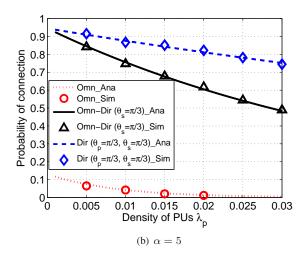
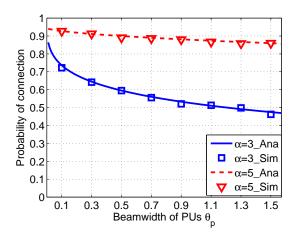


Fig. 9. Probability of connection of an SU pair versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with different α , where r=3.



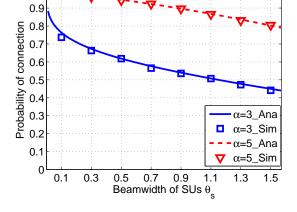


Fig. 10. Probability of connection of an SU pair versus the beamwidth of PUs with different values of α , where $\lambda_p = 0.01$, r = 3 and $\theta_s = \pi/3$.

Fig. 11. Probability of connection of an SU pair versus the beamwidth of SUs with different values of α , where $\lambda_p = 0.01$, r = 3 and $\theta_p = \pi/3$.

we fix $\theta_s = \frac{\pi}{3}$ and vary θ_p in the range of $(0, \frac{\pi}{2}]$. We observe from Fig. 10 that the probability of connection with larger path loss exponent α (e.g., $\alpha = 5$) is higher than that with smaller path loss exponent α (e.g., $\alpha = 3$). Moreover, increasing the beamwidth of PUs θ_p leads to the decreased probability of connection. This may owe to the effect that fewer SUs can have the spectrum since more SUs receive the detection preambles from PRs and the decreased topological connectivity of SUs due to the increased interference of PUs.

We further conduct the fourth set of simulations to investigate the impact of the beamwith of SUs θ_s . Fig. 11 presents the results where we fix $\theta_p = \frac{\pi}{3}$ and vary θ_s in the range of $(0, \frac{\pi}{2}]$. Fig. 11 also shows that the probability of connection with larger path loss exponent α is higher than that with smaller path loss exponent α due to the higher spectrum availability and the higher topological connectivity brought by larger path loss exponent α (as indicated in Fig. 4 and Fig. 6). Meanwhile, increasing the beamwidth of SUs θ_s leads to the decreased probability of connection as shown in Fig. 11. This is because 1) more SUs receive the detection preambles from PRs due to

the broader beamwidth of SUs and 2) the shorter transmission range of SUs due to the lower antenna gains.

V. DISCUSSIONS

In this section, we first extend our analysis by considering the transmission power efficiency in Section V-A. We then investigate the spatial throughput of SUs in Section V-B.

A. Transmission power efficiency

Generally, a directional antenna can cover the same transmission range as an omni-directional antenna with less power due to the higher antenna gain [37]–[39]. We extend the previous analysis on the probability of connection by considering transmission power efficiency of directional antennas. In particular, we assume that the received power at a node (a PU or an SU) is the same for all Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs.

We first consider the transmission power of PUs in Omn-CRNs. Take Fig. 12 as an example. We denote the transmission

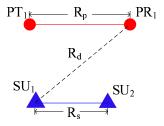


Fig. 12. Transmission range and detection range

power from PT_1 to PR_1 by P_n^o . According to the channel model given in Eq. (1), we have

$$P_r = P_p^o R_p^{-\alpha} h G_t G_r, \tag{32}$$

where R_p is the transmission range between PT_1 and PR_1 (as shown in Fig. 12), G_t and G_r are the antenna gains of PT_1 and PR₁, respectively. In Omn-CRNs, $G_t = 1$ and $G_r = 1$.

Similarly, we have the received power at a PU in Dir-CRNs as follows,

$$P_r = P_p^d R_p^{-\alpha} h G_t G_r, (33)$$

where $G_t=\frac{2\pi}{\theta_p}$ and $G_r=\frac{2\pi}{\theta_p}$ according to Eq. (2). Since the received power P_r is identical for Eq. (32) and Eq. (33) and we consider the same transmission range R_p for both Omn-CRNs and Dir-CRNs, we have the following equation after combining Eq. (32) with Eq. (33),

$$P_p^d = \frac{\theta_p^2}{4\pi^2} P_p^o.$$

Note that the transmission power of PUs in Omn-Dir-CRNs denoted by P_n^{od} is the same as Omn-CRNs. In particular, Table I summarizes the transmission power of PUs in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs (see the 2nd column).

We denote the transmission powers of SUs in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs by P_s^o , P_s^{od} and P_s^d , respectively. In a similar manner, we can obtain transmission power of SUs in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs as given in Table I (see the 3rd column).

We next denote the detection powers from a PR to an SU in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs by P_d^o , P_d^{od} and P_d^d , respectively. Following the similar analytical process, we can obtain P_d^o , P_d^{od} and P_d^d as given in Table I (see the 4th

We observe from Table I that we require less power to establish a link (either a PU link or an SU link) in Dir-CRNs than Omn-CRNs due to the higher antenna gains under the conditions of the same received power and the same transmission range. We next analyse the spectrum availability, the topological connectivity and the probability of connection with consideration of the transmission power efficiency.

Following the similar analytical procedure as presented in Section III-A, we can obtain the spectrum availability of Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs. Without too many repetitions, we omit the detailed derivation in this paper. Fig. 13 shows the probability that any SU pair has spectrum available versus node density λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with consideration of transmission power efficiency. We observe from Fig. 13 that using directional antennas at both

TABLE I ENERGY EFFICIENCY OF OMN-CRNS, OMN-DIR-CRNS AND DIR-CRNS

	Transmission power of PUs	Transmission power of SUs	Detection power
Omn-CRNs	P_p^o	P_s^o	P_d^o
Omn-Dir-CRNs	$P_p^{od} = P_p^o$	$P_s^{od} = \frac{\theta_s^2}{4\pi^2} P_s^o$	$P_d^{od} = \frac{\theta_s}{2\pi} P_d^o$
Dir-CRNs	$P_p^d = \frac{\theta_p^2}{4\pi^2} P_p^o$	$P_s^d = \frac{\theta_s^2}{4\pi^2} P_s^o$	$P_d^d = \frac{\theta_s \theta_p}{4\pi^2} P_d^o$

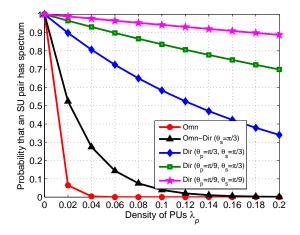
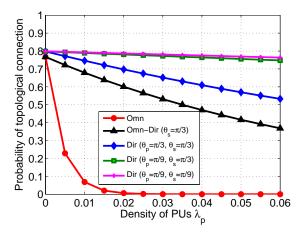


Fig. 13. Probability that any SU pair has spectrum available versus λ_p in Omn-CRNs, Dir-CRNs and Omn-Dir-CRNs with consideration of transmission power efficiency with different $F(\theta_s, \theta_p)$, where $P_d^o = 10$ and

PUs and SUs can greatly improve the spectrum availability at SUs with consideration of transmission power efficiency.



Probability of topological connection of any SU pair versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with consideration of transmission power efficiency, where $P_d^o=10$, $P_p^o=8$, $P_s^o=6$, r=3, $\lambda_s=0.0002$, $\eta=0.05$, $\sigma^2=0.01$, $\alpha=3$ and $\delta=5$.

Similar to Section III-B, we can obtain the topological connectivity of Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs. Fig. 14 shows the analytical results of the topological connectivity of Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with consideration of transmission power efficiency. Similarly, we can find that Dir-CRNs have the higher topological connectivity than Omn-CRNs and Omn-Dir-CRNs.

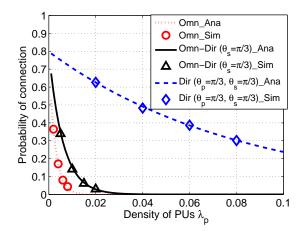


Fig. 15. Probability of connection of an SU pair versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs, where $P_d^o=10$, $P_p^o=8$, $P_s^o=6$, $\lambda_s=0.0002$, r=3, $\eta=0.05$, $\sigma^2=0.01$, $\alpha=3$ and $\delta=5$.

Finally, we also conduct a new set of simulations to investigate the impact of transmission power efficiency on the probability of connection. Fig. 15 shows the simulation results. It is shown in Fig. 15 that Dir-CRNs have much higher probability of connection than Omn-CRNs and Omn-Dir-CRNs with consideration of transmission power efficiency. This improvement mainly owes to 1) the *higher spectrum availability* and 2) the *higher topological connectivity*.

B. Throughput capacity of SUs

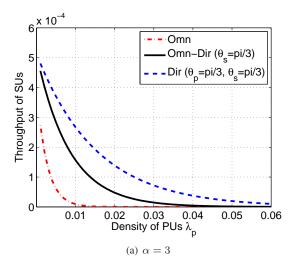
Our results also give an implication on optimizing the network throughput of SUs. In particular, following the definition of the spatial throughput of SUs as defined in [40], we have

$$\mathcal{T}_s = p_{con}\lambda_s \log(1+\delta), \tag{34}$$

where p_{con} is the probability of connection as derived in Section III, λ_s is the node density of SUs and δ is the SINR threshold.

From Eq. (34), we can conclude that the spatial throughput of SUs depends on the probability of connection p_{con} , which essentially depends on the density of SUs λ_s , the density of PUs λ_p , the beamwidth of SU θ_p , the beamwidth of θ_s and other factors (like the path loss exponent α) according to the analysis of the probability of connection given in Section III, We next investigate the impact of the density of PUs λ_p on the throughput of SUs \mathcal{T}_s .

Fig. 16 shows the throughput of SUs \mathcal{T}_s versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs. We observe that the throughput of SUs of Dir-CRNs, Omn-Dir-CRNs and Omn-CRNs decreases with the increased node density of PUs. This indicates that the throughput of SUs is significantly affected by the activities of PUs. Moreover, the throughput of SUs is also affected by the path loss component α . Furthermore, we also find that Dir-CRNs outperform Omn-Dir-CRNs and Omn-CRNs in terms of the throughput of SUs.



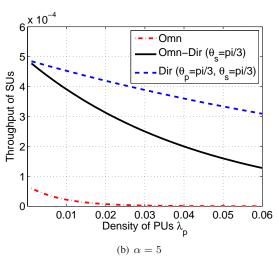


Fig. 16. Spatial throughput of SUs (bps/Hz/unit-area) versus the density of PUs λ_p in Omn-CRNs, Omn-Dir-CRNs and Dir-CRNs with different α , where $P_d=10,\ P_p=8,\ P_s=6,\ \lambda_s=0.0002,\ r=3,\ \eta=0.05,$ $\sigma^2=0.01,\ \delta=5.$

Fig. 17 shows the throughput of SUs \mathcal{T}_s versus the beamwidth of PUs θ_p in Dir-CRNs. We observe that increasing beamwidth θ_p leads to the lower throughput of SUs. Meanwhile, the throughput with larger path loss exponent (e.g., $\alpha=5$) is higher than that with smaller path loss exponent (e.g., $\alpha=3$).

We next investigate the impact of beamwidth of SUs. Fig. 18 shows the throughput of SUs \mathcal{T}_s versus the beamwidth of SUs θ_s in Dir-CRNs. We have the similar findings to those in Fig. 17. Specifically, increasing beamwidth θ_s also results in the lower throughput of SUs.

VI. CONCLUSION

In this paper, we analyse the connectivity of SUs in underlay cognitive radio networks with directional antennas (named as Dir-CRNs). Our model takes both the spectrum availability of SUs and the topological connectivity of SUs into account. Extensive simulation results verify the accuracy of our proposed model. In conclusion, this paper provides the following major findings:

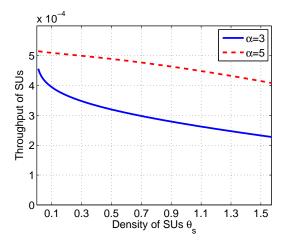


Fig. 18. Spatial throughput of SUs (bps/Hz/unit-area) versus the beamwidth of SUs with different values of α , where $P_d=10$, $P_p=8$, $P_s=6$, $\lambda_p=0.01$, $\lambda_s=0.0002$, $\eta=0.05$, $\sigma^2=0.01$, $\delta=5$, r=3 and $\theta_p=\pi/3$

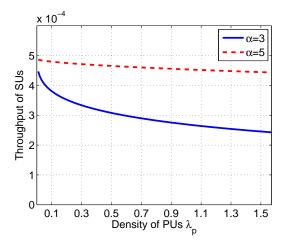


Fig. 17. Spatial throughput of SUs (bps/Hz/unit-area) versus the beamwidth of PUs with different values of α , where $P_d=10$, $P_p=8$, $P_s=6$, $\lambda_p=0.01$, $\lambda_s=0.0002$, $\eta=0.05$, $\sigma^2=0.01$, $\delta=5$, r=3 and $\theta_s=\pi/3$.

- The connectivity of SUs heavily depends on the *spectrum* availability and the *topological connectivity*.
- Dir-CRNs have the higher connectivity than Omn-CRNs.
 This improvement mainly owes to the lower interference caused by PUs, the higher spectrum availability and the higher topological connectivity, both of which are brought by directional antennas.
- Dir-CRNs have the higher transmission power efficiency than Omn-CRNs. In particular, it requires less transmission power to establish a PU link or an SU link in Dir-CRNs than that in Omn-CRNs.
- Dir-CRNs have the higher throughput of SUs than Omn-CRNs due to the lower interference, the higher spectrum availability and the higher topological connectivity.

Our proposed Dir-CRNs can potentially improve the performance of SUs in practical scenarios, such as smart grids [41], [42], vehicular ad hoc networks [43] and mobile crowdsensing [44], [45].

APPENDIX A

Proof of Lemma 1

Proof: Recall the fact that STs are distributed according to a homogenous PPP with density λ_s . We then follow the derivation similar to [11] and have the result that active STs under the condition of an SU pair with the spectrum in Dir-CRNs and in Omn-CRNs also follow a thinning homogeneous PPP with density $\lambda_s \cdot p_{ij}^d$ and density $\lambda_s \cdot p_{ij}^o$, respectively, where p_{ij}^d and p_{ij}^o are given by Eq. (5) and Eq. (10), respectively. We then have the results given in Lemma 1.

REFERENCES

- A. Ahmad, S. Ahmad, M. H. Rehmani, and N. U. Hassan, "A survey on radio resource allocation in cognitive radio sensor networks," *IEEE Communications Surveys Tutorials*, vol. 17, no. 2, pp. 888–917, Secondquarter 2015.
- [2] H. Sun and A. Nallanathan and C. X. Wang and Y. Chen, "Wideband spectrum sensing for cognitive radio networks: a survey," *IEEE Wireless Communications*, vol. 20, no. 2, pp. 74–81, April 2013.
- Communications, vol. 20, no. 2, pp. 74–81, April 2013.
 E. Z. Tragos, S. Zeadally, A. G. Fragkiadakis, and V. A. Siris, "Spectrum assignment in cognitive radio networks: A comprehensive survey," *IEEE Communications Surveys Tutorials*, vol. 15, no. 3, pp. 1108–1135, Third 2013.
- [4] A. Goldsmith, S. Jafar, I. Maric, and S. Srinivasa, "Breaking spectrum gridlock with cognitive radios: An information theoretic perspective," *Proceedings of the IEEE*, vol. 97, no. 5, pp. 894–914, May 2009.
- [5] L. Sibomana, H. J. Zepernick, H. Tran, and C. Kabiri, "A framework for packet delay analysis of point-to-multipoint underlay cognitive radio networks," *IEEE Transactions on Mobile Computing*, vol. PP, no. 99, pp. 1–1, 2016.
- [6] A. Kaushik, S. K. Sharma, S. Chatzinotas, B. Ottersten, and F. K. Jondral, "On the performance analysis of underlay cognitive radio systems: A deployment perspective," *IEEE Transactions on Cognitive Communications and Networking*, vol. 2, no. 3, pp. 273–287, 2016.
- [7] M. Matinmikko, H. Okkonen, M. Palola, S. Yrjola, P. Ahokangas, and M. Mustonen, "Spectrum sharing using licensed shared access: the concept and its workflow for lte-advanced networks," *IEEE Wireless Communications*, vol. 21, no. 2, pp. 72–79, April 2014.
- [8] J. Liu, Q. Zhang, Y. Zhang, Z. Wei, and S. Ma, "Connectivity of two nodes in cognitive radio ad hoc networks," in *Proc. IEEE WCNC*, 2013.
- [9] M. T. Masonta, M. Mzyece, and N. Ntlatlapa, "Spectrum decision in cognitive radio networks: A survey," *IEEE Communications Surveys Tutorials*, vol. 15, no. 3, pp. 1088–1107, Third 2013.
- [10] D. Zhai, M. Sheng, X. Wang, and Y. Zhang, "Local connectivity of cognitive radio ad hoc networks," in 2014 IEEE Global Communications Conference, Dec 2014, pp. 1078–1083.
- [11] D. Liu, E. Liu, Z. Zhang, R. Wang, Y. Ren, Y. Liu, I. W. H. Ho, X. Yin, and F. Liu, "Secondary network connectivity of ad hoc cognitive radio networks," *IEEE Communications Letters*, vol. 18, no. 12, pp. 2177–2180, Dec 2014.
- [12] Y. Liu, Y. Cui, and X. Wang, "Connectivity and transmission delay in large-scale cognitive radio ad hoc networks with unreliable secondary links," *IEEE Transactions on Wireless Communications*, vol. 14, no. 12, pp. 7016–7029, Dec 2015.
- [13] Y. Liu, J. Gao, J. Yu, and C. Yin, "Local connectivity for heterogeneous overlaid wireless networks," Ad Hoc Networks, 2016.
- [14] W. Guo and X. Huang, "Multicast communications in cognitive radio networks using directional antennas," Wireless Communications and Mobile Computing, vol. 15, no. 2, pp. 260–275, 2015.
- [15] Y. Dai, J. Wu, and Y. Zhao, "Boundary helps: Reliable route selection with directional antennas in cognitive radio networks," *IEEE Transac*tions on Vehicular Technology, vol. 64, no. 9, pp. 4135–4143, 2015.
- [16] L. T. Dung and B. An, "A modeling framework for supporting and evaluating connectivity in cognitive radio ad hoc networks with beamforming," Wireless Networks, pp. 1–13, 2016. [Online]. Available: http://dx.doi.org/10.1007/s11276-016-1252-9
- [17] H. Yazdani and A. Vosoughi, "On cognitive radio systems with directional antennas and imperfect spectrum sensing," in 2017 IEEE International Conference on Acoustics, Speech and Signal Processing (ICASSP), 2017, pp. 3589–3593.

- [18] M. Ni, L. Zheng, F. Tong, J. Pan, and L. Cai, "A geometrical-based throughput bound analysis for device-to-device communications in cellular networks," *IEEE Journal on Selected Areas in Communications*, vol. 33, no. 1, pp. 100–110, Jan 2015.
- [19] J. Qiao, X. Shen, J. Mark, Q. Shen, Y. He, and L. Lei, "Enabling device-to-device communications in millimeter-wave 5g cellular networks," *IEEE Communications Magazine*, vol. 53, no. 1, pp. 209–215, January 2015.
- [20] J. Deng, O. Tirkkonen, R. Freij-Hollanti, T. Chen, and N. Nikaein, "Resource allocation and interference management for opportunistic relaying in integrated mmwave/sub-6 ghz 5g networks," *IEEE Com*munications Magazine, vol. 55, no. 6, pp. 94–101, 2017.
- [21] J. F. C.Kingman, *Poission Processes*. Clarendon Press: Oxford, 1993.
- [22] J. G. Andrews, F. Baccelli, and R. K. Ganti, "A tractable approach to coverage and rate in cellular networks," *IEEE Transactions on Communications*, vol. 59, no. 11, pp. 3122–3134, November 2011.
- [23] O. Georgiou, S. Wang, M. Z. Bocus, C. P. Dettmann, and J. P. Coon, "Directional antennas improve the link-connectivity of interference limited ad hoc networks," in *IEEE 26th Annual International Symposium* on *Personal, Indoor, and Mobile Radio Communications (PIMRC)*, Aug 2015, pp. 1311–1316.
- [24] T. S. Rappaport, Wireless communications: principles and practice, 2nd ed. Upper Saddle River, N.J.: Prentice Hall PTR, 2002.
- [25] H.-N. Dai, K.-W. Ng, and M.-Y. Wu, "On busy-tone based MAC protocol for wireless networks with directional antennas," Wireless Personal Communications, vol. 73, no. 3, pp. 611 – 636, 2013.
- [26] M. D. Renzo, "Stochastic geometry modeling and performance evaluation of mmwave cellular communications," in 2015 IEEE International Conference on Communications (ICC), June 2015, pp. 5992–5997.
- [27] G. Zhang, Y. Xu, X. Wang, and M. Guizani, "Capacity of hybrid wireless networks with directional antenna and delay constraint," *IEEE Transactions on Communications*, vol. 58, no. 7, pp. 2097–2106, 2010.
- [28] T. Nitsche, C. Cordeiro, A. Flores, E. W. Knightly, E. Perahia, and J. C. Widmer, "IEEE 802.11ad: directional 60 GHz communication for multi-Gigabit-per-second Wi-Fi," *IEEE Communications Magazine*, vol. 52, no. 12, pp. 132–141, December 2014.
- [29] H.-N. Dai and Q. Zhao, "On the delay reduction of wireless ad hoc networks with directional antennas," EURASIP Journal on Wireless Communications and Networking, vol. 2015, no. 1, pp. 1–13, 2015.
- [30] M. Takai, J. Martin, R. Bagrodia, and A. Ren, "Directional virtual carrier sensing for directional antennas in mobile ad hoc networks," in *Proceedings of ACM MobiHoc*, 2002.
- [31] C. A. Balanis, Antenna Theory: Analysis and Design, 3rd ed. New York: John Wiley & Sons, 2005.

- [32] A. Ghasemi and E. S. Sousa, "Spectrum sensing in cognitive radio networks: requirements, challenges and design trade-offs," *IEEE Communications Magazine*, vol. 46, no. 4, pp. 32–39, April 2008.
- [33] A. Rabbachin, T. Q. S. Quek, H. Shin, and M. Z. Win, "Cognitive network interference," *IEEE Journal on Selected Areas in Communications*, vol. 29, no. 2, pp. 480–493, February 2011.
- [34] H.-N. Dai, K.-W. Ng, M. Li, and M.-Y. Wu, "An Overview of Using Directional Antennas in Wireless Networks," *International Journal of Communication Systems (Wiley)*, vol. 26, no. 4, pp. 413 – 448, 2013.
- [35] T. Bai and R. Heath, "Coverage and rate analysis for millimeter-wave cellular networks," *IEEE Transactions on Wireless Communications*, vol. 14, no. 2, pp. 1100–1114, Feb 2015.
- [36] C. Bettstetter, "On the Connectivity of Ad Hoc Networks," The Computer Journal, vol. 47, no. 4, pp. 432 447, 2004.
- [37] Q. Wang, H.-N. Dai, Z. Zheng, M. Imran, and A. V. Vasilakos, "On Connectivity of Wireless Sensor Networks with Directional Antennas," *Sensors*, vol. 17, no. 1, 2017.
- [38] X. Zhou, S. Durrani, and H. Jones, "Connectivity Analysis of Wireless Ad Hoc Networks with Beamforming," *IEEE Transactions on Vehicular Technology*, vol. 58, no. 9, pp. 5247 – 5257, 2009.
- [39] J. Qiao, X. Shen, J. W. Mark, Q. Shen, Y. He, and L. Lei, "Enabling device-to-device communications in millimeter-wave 5g cellular networks," *IEEE Communications Magazine*, vol. 53, no. 1, pp. 209–215, January 2015.
- [40] S. Lee, R. Zhang, and K. Huang, "Opportunistic wireless energy harvesting in cognitive radio networks," *IEEE Transactions on Wireless Communications*, vol. 12, no. 9, pp. 4788–4799, 2013.
- [41] Y. Zhang, R. Yu, M. Nekovee, Y. Liu, S. Xie, and S. Gjessing, "Cognitive machine-to-machine communications: visions and potentials for the smart grid," *IEEE Network*, vol. 26, no. 3, pp. 6–13, May 2012.
- [42] R. Yu, W. Zhong, S. Xie, C. Yuen, S. Gjessing, and Y. Zhang, "Balancing power demand through ev mobility in vehicle-to-grid mobile energy networks," *IEEE Transactions on Industrial Informatics*, vol. 12, no. 1, pp. 79–90, 2016.
- [43] X. He, H. Zhang, T. Luo, and W. Shi, "Network capacity analysis for cellular based cognitive radio VANET in urban grid scenario," *J. Comm. Inform. Networks*, vol. 2, no. 2, pp. 136–146, 2017.
- [44] G. Yang, S. He, Z. Shi, and J. Chen, "Promoting cooperation by the social incentive mechanism in mobile crowdsensing," *IEEE Communi*cations Magazine, vol. 55, no. 3, pp. 86–92, 2017.
- [45] G. Yang, S. He, and Z. Shi, "Leveraging crowdsourcing for efficient malicious users detection in large-scale social networks," *IEEE Internet* of Things Journal, vol. 4, no. 2, pp. 330–339, 2017.